Coherent-Feedback Formulation of a Continuous Quantum Error Correction Protocol

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Talk plan

- Reminder of quantum error correction (QEC).
- Continuous QEC.
- Description of a scheme for coherent-feedback QEC for a simple 3 qubit bit flip code.
- Concluding remarks.

Quantum error correction (QEC)

- Quantum error correction is essential for quantum information processing since qubits are susceptible to decoherence.
- When decoherence alters the state of a qubit, a QEC algorithm acts to restore it to the state prior to decoherence.

QEC principles

- Main ingredient is to introduce reduncancy by encoding a logical qubit into a number of physical qubits.
- Simple 3 qubit code: A *logical* qubit is encoded by 3 *physical* qubits. Logical qubit |0 >_L encoded by 3 physical qubits |000 > and |1 >_L encoded by |111 >.
- A logical state a |0 >_L + b |1 >_L is encoded as a |000 > + b | 111 >. The 3 qubit codespace is C = span{|000 >, |111>}.

The 3 qubit bit flip code

- This simple code belongs to a class of QEC codes called *stabilizer codes* (Gottesman, PRA 54, 1862)
- C = span{| 000 >, |111 >}
- Correctable errors are single qubit bit flips X₁, X₂ or X₃
- Error is determined by measuring the parity Z₁Z₂ between qubits 1 and 2, and Z₂Z₃ between qubits 2 and 3. Measurement results called the error syndrome.
- Error syndromes are: {1,1} (no error), {-1, 1} (qubit 1 has flipped), {1,-1} (qubit 3 has flipped) and {-1,-1} (qubit 2 has flipped)

Continuous QEC

- Most proposed QEC schemes are discrete. Error detection and recovery operations are done periodically with a sufficiently small period.
- In continuous QEC, the idea is to detect and correct errors continuously as they occur. Suitable for low level continuous time differential equation based models.
- Continuous QEC using continuous monitoring and measurement-feedback: Ahn, Doherty & Landahl et al, PRA 65, 042301; Ahn, Wiseman & Milburn, PRA 65, 042301; Chase, Landahl & Geremia, PRA 77, 032304.

Why coherent-feedback?

- Not necessary to go up to the "macroscopic" level and have interfaces to electronic circuits for measurements. Not limited by bandwidth of electronic devices.
- No classical processing required and avoids challenges imposed by the requirement of such processing; e.g., numerical integration of nonlinear quantum filtering equations in real-time.
- Entirely "on-chip" implementation; a controller can be on the same hardware platform as the controlled quantum system. In particular, in solid state monolithic circuit QED.

Quantum network notation and operations

- Open Markov quantum system G = (S,L,H).
- Concatenation product $G_2 \boxplus G_1 = (S_2, L_2, H_2) \boxplus (S_1, L_1, H_1) = \left(\begin{bmatrix} S_2 & 0 \\ 0 & S_1 \end{bmatrix}, \begin{bmatrix} L_2 \\ L_1 \end{bmatrix}, H_1 + H_2 \right)$
- Series product

$$G_2 \triangleleft G_1 = (S_2, L_2, H_2) \triangleleft (S_1, L_1, H_1)$$

= $(S_2S_1, S_2L_1 + L_2, H_1 + H_2 + \Im\{L_2^{\dagger}S_2L_1\})$

Gough & James, IEEE-TAC (to appear), 2009, arXiv:0708.4483

Continuous parity measurement



Kerckhoff, Bouten, Silberfarb & Mabuchi, PRA 79, 024305

$dU(t) = \left((\sqrt{\kappa}b_1 + \sqrt{\kappa}b_2 + \alpha) dA^*(t) - h.c. \right)$
$-\frac{(\sqrt{\kappa}b_1+\sqrt{\kappa}b_2+\alpha)^*(\sqrt{\kappa}b_1+\sqrt{\kappa}b_2+\alpha)}{dt}dt$
2
$+\sum_{k=1}^{2} \left(\sqrt{\gamma}\sigma^{(i)}dB^{*}(t) - h.c \frac{\gamma}{2}\sigma^{*}\sigma dt\right)$
$+\frac{\kappa}{2}(b_1^*b_2 - h.c.)dt + g\sum_{i=1}^2 (\sigma^{(i)*}b_i - h.c.)dt$
$+\left(\frac{\bar{\alpha}\sqrt{\kappa}}{2}(b_1+b_2)-h.c.\right)dt\Big)U(t)$
$d\overline{U}(t) = ((\Pi_{12} - I)d\Lambda(t) + \alpha \Pi_{12}dA^*(t) - \overline{\alpha}dA(t))$
$-rac{ lpha ^2}{2}dtig)ar{U}(t)$
$\lim_{\alpha, \kappa \to \infty} \lim_{g \to \infty} \ (U(t) - \overline{U}(t))\psi \ = 0 \text{ for all } \psi$
$\frac{\alpha}{\sqrt{\kappa}} = \text{const}$
in the reduced Hilbert space.

But cannot include bit flip errors!

Modified parity measurement model

- Full single atom-cavity-field model with bit flip errors: $dU(t) = \left((\sqrt{\kappa}b + \alpha)dA_1^*(t) - h.c. - \frac{1}{2}(\sqrt{\kappa}b + \alpha)^*(\sqrt{\kappa}b + \alpha)dt + \sqrt{\Gamma}\sigma_X dA_2^*(t) - h.c. - \frac{\Gamma}{2}\sigma_X^*\sigma_X dt + \sqrt{\gamma}\sigma dA_3^*(t) - h.c. - \frac{\gamma}{2}\sigma^*\sigma dt + g(\sigma^*b - h.c.)dt \right) U(t)$
- Reduced atom-field model with bit flip errors:

$$d\bar{U}(t) = \left((Z-I)d\Lambda_{11}(t) + \alpha Z dA_1^*(t) - \bar{\alpha} dA_1(t) \right.$$
$$\left. + \sqrt{\Gamma} X dA_2^*(t) - h.c. - \frac{1}{2} (|\alpha|^2 + \gamma) dt \right) \bar{U}(t)$$

 $\lim_{\substack{g, \kappa \to \infty \\ \frac{g}{\kappa} = \text{ const}} \| (U(t) - \overline{U}(t)) \psi \| \text{ for all } \psi \text{ in the reduced Hilbert space.}$

Modified parity measurement model

• The reduced model can be written as: $\left(\begin{bmatrix} Z & 0 \\ 0 & 1 \end{bmatrix}, \begin{bmatrix} 0 \\ \sqrt{\Gamma}X \end{bmatrix}, 0 \right) \triangleleft \left(I, \begin{bmatrix} \alpha \\ 0 \end{bmatrix}, 0 \right)$

• Reduced model for two coherently driven cavities:

$$\begin{pmatrix} \begin{bmatrix} Z_2 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix}, \begin{bmatrix} 0 \\ 0 \\ \sqrt{\Gamma}X_2 \end{bmatrix}, 0 \land d \begin{pmatrix} \begin{bmatrix} Z_1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix}, \begin{bmatrix} 0 \\ \sqrt{\Gamma}X_1 \\ 0 \end{bmatrix}, 0 \land d \begin{pmatrix} I, \begin{bmatrix} \alpha \\ 0 \\ 0 \end{bmatrix}, 0 \end{pmatrix}$$
$$= \begin{pmatrix} \begin{bmatrix} Z_2Z_1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix}, \begin{bmatrix} \alpha Z_2Z_1 \\ \sqrt{\Gamma}X_1 \\ \sqrt{\Gamma}X_2 \end{bmatrix}, 0 \end{pmatrix}$$

The "bare bones" of it



Coherent lasers drive atomic Raman transitions between the two ground states of the atom to correct bit flips.
To make it work, need more than this ...

The actual scheme



• Quantum switches R1, R2 inserted to facilitate switching to higher amplitude bit flip correcting Raman lasers.

Qubit atomic level scheme



Raman transitions with ac Stark shift compensation

ac Stark shift compensation • Given by the terms: $\left(1, \sqrt{\gamma} \left(\sigma_{gr} + \sigma_{gG}\right), \Delta \left(\frac{1}{2}\Pi_r - \Pi_G\right)\right)$ Raman transitions Raman transitions $\boxplus \left(1, \sqrt{\gamma} \left(\sigma_{hr} + \sigma_{hH} \right), \Delta \left(\frac{1}{2} \Pi_r - \Pi_H \right) \right)$ • Together with coupling to probe laser becomes: $\left(\left| \begin{array}{cc} Z & 0 \\ 0 & 1 \end{array} \right|, 0, 0 \right) \boxplus \left(1, \sqrt{\gamma} \left(\sigma_{gr} + \sigma_{gG} \right), \Delta \left(\frac{1}{2} \Pi_r - \Pi_G \right) \right) \right.$ $\boxplus \left(1, \sqrt{\gamma} \left(\sigma_{hr} + \sigma_{hH}\right), \Delta \left(\frac{1}{2}\Pi_r - \Pi_H\right)\right)$

QSDE model for switch/relay



• Simple QSDE model:

 $\left(\begin{bmatrix} \Pi_g & -\Pi_h \\ -\Pi_h & \Pi_g \end{bmatrix}, \begin{bmatrix} \beta \Pi_g \\ -\beta \Pi_h \end{bmatrix}, 0 \right) \boxplus \left(\begin{bmatrix} \Pi_h & -\sigma_{hg} \\ -\sigma_{gh} & \Pi_g \end{bmatrix}, 0, 0 \right)$

Detailed physical model: H. Mabuchi, arXiv:0907.2720

QEC network description



Half network $G_p \boxplus G_f$ (p=probe path, f=feedback path): $G_p = R_{12} \triangleleft B_3 \triangleleft ((Q_{13} \triangleleft Q_{21}) \boxplus (1,0,0)) \triangleleft B_1,$ $G_f = (Q_{11} \boxplus Q_{32} \boxplus Q_{22}) \triangleleft (B_5 \boxplus_2 (1,0,0))$ $\triangleleft (R_{11} \boxplus (1,0,0))$

QEC network master equation

• The QEC network master equation in the limit that $\Delta, \beta \rightarrow \infty$ with β^2/Δ constant is:

$$\dot{\rho}_t = -i[H, \rho_t] + \sum_{i=1}^7 \left(L_i \rho_t L_i^* - \frac{1}{2} \{ L_i^* L_i, \rho_t \} \right)$$

where $(\Omega \equiv \beta^2 / \gamma \Delta)$,

$$H = \sqrt{2}\Omega\Pi_{g}^{(R_{1})}\Pi_{h}^{(R_{2})}X_{1} + \sqrt{2}\Omega\Pi_{h}^{(R_{1})}\Pi_{g}^{(R_{2})}X_{3}$$
$$-\Omega\Pi_{g}^{(R_{1})}\Pi_{g}^{(R_{2})}X_{2},$$
$$L_{1} = \frac{\alpha}{\sqrt{2}} \left\{ \sigma_{hg}^{(R_{1})}(1 + Z_{1}Z_{2}) + \Pi_{h}^{(R_{1})}(1 - Z_{1}Z_{2}) \right\}$$
$$L_{2} = \frac{\alpha}{\sqrt{2}} \left\{ \sigma_{gh}^{(R_{1})}(1 - Z_{1}Z_{2}) + \Pi_{g}^{(R_{1})}(1 + Z_{1}Z_{2}) \right\}$$
$$L_{3} = \frac{\alpha}{\sqrt{2}} \left\{ \sigma_{hg}^{(R_{2})}(1 + Z_{3}Z_{2}) + \Pi_{h}^{(R_{2})}(1 - Z_{3}Z_{2}) \right\}$$
$$L_{4} = \frac{\alpha}{\sqrt{2}} \left\{ \sigma_{gh}^{(R_{2})}(1 - Z_{3}Z_{2}) + \Pi_{g}^{(R_{2})}(1 + Z_{3}Z_{2}) \right\}$$
$$L_{5} = \sqrt{\Gamma}X_{1}, \quad L_{6} = \sqrt{\Gamma}X_{2}, \quad L_{7} = \sqrt{\Gamma}X_{3}.$$

Ideal QEC network performance



Fidelity = $\langle \psi_0 | \rho(t) | \psi_0 \rangle$; Simulation parameters: $\Gamma = 0.01$, $\alpha = 10$, $|\psi_0 \rangle = (| ggg \rangle - |hhh \rangle)/\sqrt{2}$; $\rho(0) = |\psi_0 \rangle \langle \psi_0 |$.

Ideal QEC network performance



Fidelity = $\langle \psi_0 | \rho(T) | \psi_0 \rangle$; Simulation parameters: $\Gamma = 0.01$, $\alpha = 10, \beta = 30$, $|\psi_0 \rangle = a |ggg \rangle + \int (1-a^2)e^{i\theta} | hhh \rangle$, a in [-1,1], θ in [- π,π]; $\rho(0) = |\psi_0 \rangle \langle \psi_0 |$.

Concluding remarks

- We have proposed a coherent-feedback QEC scheme for a simple 3 qubit QEC code that protects against single bit flip errors; can be easily adapted to a 3 qubit phase flip code.
- Simulations of the QEC network master equation indicates the scheme can slow down decoherence due to single bit flips.
- Ideas for the future: Adaptation to more complex stabilizer codes, but necessarily also with more complex quantum circuits. Perhaps also to non-stabilizer codes (more challenging?).